

CC1352P SimpleLink™ Wireless MCU Device Revision E



ABSTRACT

This document describes the known exceptions to functional specifications (advisories) to the CC1352P SimpleLink™ device.

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1 Advisories Matrix

Table 1-1 lists all advisories, modules affected, and the applicable silicon revisions.

Table 1-1. Advisories Matrix

MODULE	DESCRIPTION	SILICON REVISIONS AFFECTED
		E
Radio	Advisory Radio_01 — Proprietary radio modes: spurious emissions can affect regulatory compliance	Yes
Radio	Advisory Radio_02 — High-power PA operation at temperatures below -20°C may affect the 32-kHz crystal oscillator	Yes
Power	Advisory Power_03 — Increased voltage ripple at low supply voltages when DC/DC converter is enabled	Yes
PKA	Advisory PKA_01 — Public key accelerator (PKA) interrupt line is always high when module is enabled and PKA is idle	Yes
PKA	Advisory PKA_02 — Public key accelerator (PKA) RAM is not byte accessible	Yes
I2C	Advisory I2C_01 — I ² C module master status bit is set late	Yes
I2S	Advisory I2S_01 — I ² S bus faults are not reported	Yes
CPU	Advisory CPU_01 — Arm® Errata #838869: Store immediate overlapping exception return operation might vector to incorrect interrupt	Yes
CPU	Advisory CPU_02 — Arm® Errata #752770: Interrupted loads to SP can cause erroneous behavior	Yes
CPU	Advisory CPU_03 — Arm® Errata #776924 VDIV or VSQRT instructions might not complete correctly when very short ISRs are used	Yes
CPU, System	Advisory CPU_Sys_01 — The SysTick calibration value (register field CPU_SCS.STCR.TENMS) used to set up 10-ms periodic ticks is incorrect when the system CPU is running off divided down 48-MHz clock	Yes
System	Advisory Sys_01 — Device might boot into ROM serial bootloader when waking up from shutdown	Yes
System Controller	Advisory SYSCTRL_01 — Resets occurring in a specific 2-MHz period during initial power up are incorrectly reported	Yes
IO Controller	Advisory IOC_01 — Limited number of DIOs available for the bootloader backdoor	Yes
SRAM	Advisory SRAM_01 — Reserved addresses within SRAM_MMR region alias into SRAM array	Yes
General-Purpose Timer	Advisory GPTM_01 — An incorrect value might be written to the general-purpose (GP) timers MMRs (memory mapped registers) when simultaneously accessing the PKA (public key accelerator) engine and/or the AES (advanced encryption standard) engine from a different master	Yes
ADC	Advisory ADC_01 — Periodic ADC trigger at 200 kHz rate can be ignored when XOSC_HF is turned on or off	Yes
ADC	Advisory ADC_02 — ADC samples can be delayed by 2 or 14 clock cycles (24 MHz) when XOSC_HF is turned on or off, resulting in sample jitter	Yes
ADC	Advisory ADC_03 — Software can hang when reading the ADC FIFO if a single manual ADC trigger is generated immediately after the ADC is enabled	Yes
ADC	Advisory ADC_04 — Misbehaving ADC FIFO status flags in the AUX_ANAIF:ADCFIFOSTAT register (OVERFLOW, FULL, ALMOST_FULL, and EMPTY)	Yes
ADC	Advisory ADC_05 — Writing any value to AUX_ANAIF:ADCTRIG.START will create an ADC trigger	Yes

2 Nomenclature, Package Symbolization, and Revision Identification

2.1 Device and Development Support-Tool Nomenclature

To designate the stages in the product development cycle, Texas Instruments™ assigns prefixes to the part numbers of all devices and support tools. Each device has one of three prefixes: X, P, or null (for example, XCC1352P). Texas Instruments recommends two of three possible prefix designators for its support tools: TMDX and TMDS. These prefixes represent evolutionary stages of product development from engineering prototypes (X/TMDX) through fully qualified production devices/tools (null/TMDS).

Device development evolutionary flow:

- X** Experimental device that is not necessarily representative of the final device's electrical specifications and may not use production assembly flow.
- P** Prototype device that is not necessarily the final silicon die and may not necessarily meet final electrical specifications.
- null** Production version of the silicon die that is fully qualified.

Support tool development evolutionary flow:

- TMDX** Development-support product that has not yet completed Texas Instruments internal qualification testing.
- TMDS** Fully-qualified development-support product.

X and P devices and TMDX development-support tools are shipped against the following disclaimer:

"Developmental product is intended for internal evaluation purposes."

Production devices and TMDS development-support tools have been characterized fully, and the quality and reliability of the device have been demonstrated fully. TI's standard warranty applies.

Predictions show that prototype devices (X or P) have a greater failure rate than the standard production devices. Texas Instruments recommends that these devices not be used in any production system because their expected end-use failure rate still is undefined. Only qualified production devices are to be used.

2.2 Devices Supported

This document supports the following device:

- [CC1352P](#)

2.3 Package Symbolization and Revision Identification

Figure 2-1 and Table 2-1 describe package symbolization and the device revision code.

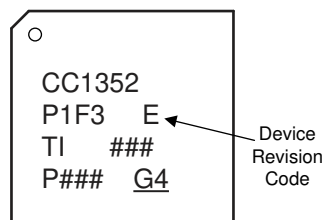


Figure 2-1. Package Symbolization for Silicon Revision E

Table 2-1. Revision Identification

DEVICE REVISION CODE	SILICON REVISION
E	PG2.1

3 Advisories

Radio_01 *Proprietary radio modes: spurious emissions can affect regulatory compliance*

Revisions Affected: Revision E and earlier

Details: When device internal load capacitors are used with the external 48-MHz crystal, energy couples from the crystal oscillator circuit to the RF output. This coupling causes spurious emissions at $N \times 48$ MHz from carrier frequency. This includes, but is not limited to, the frequency bands supported by the device covered by the following regulations:

When using the +14-dBm RF power amplifier

-
- ARIB T-108 (Japan)

When using the +20-dBm RF power amplifier

-
- 470- to 510-MHz (China) Band
- FCC CFR47 Part 15 (US)
- ETSI EN 300 220 (Europe)
- ETSI EN 300 328 (Europe)
- ETSI EN 300 440 (Europe)

Workaround: For compliance with affected standards, external load capacitors might be needed for the 48-MHz crystal to reduce spurious emissions. Internal capacitors (default 7-pF connected capacitance) must then be disconnected internally.

This workaround is implemented by defining the following symbols in the included customer configuration file (ccfg.c) available in all Software Development Kit (SDK) examples:

```
#define SET_CCFG_MODE_CONF_XOSC_CAPARRAY_DELTA -128
#define SET_CCFG_MODE_CONF_XOSC_CAP_MOD 0
```

Radio_02 *High-Power PA Operation at Temperatures Below -20°C May Affect the 32-kHz Crystal Oscillator*

Revisions Affected: Revision E and earlier

Details: When using the high-power PA at temperatures below -20°C and high output power, the PA may affect the 32-kHz crystal oscillator due to RF load impedance mismatch. In this situation the crystal oscillator will stop, and provided the clock loss detector is enabled (OSC_DIG:CTL0.CLK_LOSS_EN = 1), the device will reset. Antenna impedances outside of VSWR of 2:1 must be avoided in all operating scenarios.

Workaround:

- For applications operating below -20°C, it is of particular importance to accurately follow the reference design for the RF balun and -matching network with respect to component values and layout. Amplitude- and phase balance through the balun must be <1dB and <6 degrees, respectively.
- For PCB designs not adhering to the TI recommended guidelines or in applications where antenna impedance is subject to change due to external factors, output power should be limited to 17 dBm maximum.
- Make sure the clock loss detector is enabled, OSC_DIG:CTL0.CLK_LOSS_EN = 1, to properly reset the device should the crystal oscillator be stopped.
- Using an integrated balun instead of the discrete component reference design is expected to improve the robustness against clock loss due to impedance mismatch. Recommended integrated baluns are:

- Johanson Technology 1720BL15B0200E
- Anaren BD0826J50200AH

Power_03

Increased voltage ripple at low supply voltages when DC/DC converter is enabled

Revisions Affected: Revision E and earlier

Details: At supply voltages <2.0V, a hardware control module disables the DC/DC converter to maximize system efficiency. This module does not have enough hysteresis, causing approx 10 mV of ripple on the VDDR regulated power supply. Based on internal testing of the device, it is not anticipated that this erratum affects RF performance. However, these test results cannot ensure that a customer's application or end equipment will not be affected.

Workaround: Use the TI-provided Power driver (PowerCC26X2.c) which automatically disables the DC/DC converter when supply voltage is <2.2V.

The workaround is available in all SDK versions.

PKA_01 ***Public key accelerator (PKA) interrupt line is always high when module is enabled and PKA is idle***

Revisions Affected: Revision E and earlier

Details: When the PKA module is enabled and idle, the interrupt line is always high and the interrupt can thus not be used as is.

Workaround: The workaround is to disable the PKA interrupt in the interrupt service routine while the PKA module is idle and re-enable the interrupt right after starting an operation.

The workaround is implemented in the TI-provided cryptography drivers (ECDHCC26X2.c, ECDSACC26X2.c, ECJPAKECC26X2.c_list.c) in the following SimpleLink software development kit (SDK) versions:

- SimpleLink CC13x2 SDK 1.60.00.xx and later

PKA_02 ***Public key accelerator (PKA) RAM is not byte accessible***

Revisions Affected: Revision E and earlier

Details: When accessing the PKA RAM, the RAM is not byte accessible. If a single byte is accessed (read or written), 4 bytes will be accessed instead.

Workaround: The workaround is to use word access (4 bytes) when accessing the PKA RAM.

The workaround is implemented in the TI-provided cryptography drivers (ECDHCC26X2.c, ECDSACC26X2.c, ECJPAKECC26X2.c_list.c) in the following SimpleLink software development kit (SDK) versions:

- SimpleLink CC13x2 SDK 1.60.00.xx and later

I2C_01 *I²C module master status bit is set late*

Revisions Affected: Revision E and earlier

Details: The I2C.MSTAT[0] bit is not set immediately after writing to the I2C.MCTRL register. This can lead an I²C master to believe it is no longer busy and continuing to write data.

Workaround: Add four NOPs between writing to the MCTRL register and polling the MSTAT register.
The workaround is implemented in the TI-provided I2C Master driver (I2CCC26XX.c) and in the I2C driver Library APIs (driverlib/i2c.c).
The workaround is available in all Software Development Kit (SDK) versions.

I2S_01 *I²S bus faults are not reported*

Revisions Affected: Revision E and earlier

Details: The I²S module will not set the bus error interrupt flag (I2S0.IRQFLAGS.BUS_ERR) if an I²S read or write causes a system bus fault that results from access to illegal addresses (usage error).

Workaround: Software must ensure that memory area used by the I²S DMA is accessible, meaning that the memory is powered on and the system bus is connected..
As an example; The TI-provided SPI driver SPIC26X2DMA.c will ensure that the flash memory is kept accessible also in Idle power mode if the transmit buffer address starts with 0x0 to ensure no bus faults occur. A similar approach needs to be taken if writing a peripheral driver utilizing I2S.

CPU_01**Arm® Errata #838869: Store immediate overlapping exception return operation might vector to incorrect interrupt****Revisions Affected:** Revision E and earlier**Details:****Configurations Affected:**

This erratum only affects systems where writeable memory locations can exhibit more than one wait state (system SRAM does not have wait states).

The Arm® Cortex®-M4 processor includes a write buffer that permits execution to continue while a store is waiting on the bus. Under specific timing conditions, during an exception return while this buffer is still in use by a store instruction, a late change in selection of the next interrupt to be taken might result in a mismatch between the interrupt acknowledged by the interrupt controller and the vector fetched by the processor.

Conditions:

- The handler for interrupt A is being executed.
- Interrupt B, of the same or lower priority than interrupt A, is pending.
- A store with immediate offset instruction is executed to a bufferable location:


```
STR/STRH/STRB <Rt>, [<Rn>, #imm]
STR/STRH/STRB <Rt>, [<Rn>, #imm]!
STR/STRH/STRB <Rt>, [<Rn>, #imm]
```
- Any number of additional data-processing instructions can be executed.
- A BX instruction is executed that causes an exception return.
- The store data has wait states applied to it such that the data is accepted at least two cycles after the BX is executed.
 - Minimally this is two cycles if the store and the BX instruction have no additional instructions between them.
 - The number of wait states required to observe this erratum needs to be increased by the number of cycles between the store and the interrupt service routine exit instruction.
- Before the bus accepts the buffered store data, another interrupt C is asserted which has the same or lower priority as A, but a greater priority than B.

Implications:

The processor should execute interrupt handler C, and on completion of handler C the processor should execute the handler for B. If the previously listed conditions are met, then this erratum results in the processor erroneously clearing the pending state of interrupt C, and then twice executing the handler for B. The first time the handler for B is executed it will be at the priority level for interrupt C. If interrupt C is pending by a level-based interrupt that is cleared by C's handler then interrupt C will be pending again after the handler for B has completed and the handler for C will be executed. If interrupt C is level based, then this interrupt will eventually become re-pending and subsequently be handled. If interrupt C is a single pulse interrupt, there is a possibility that this interrupt will be lost.

This bug is triggered in a rare condition. In cases where STORE experiences more than 2 wait cycles, workarounds must be used by the software developer.

- This erratum does not apply for TI-RTOS interrupts, which ensures that no *store with immediate offset* occurs within the last 5 instructions of the interrupt routine. See the following files included in all SDKs for further implementation details:
 - kernel/tirtos/packages/ti/sysbios/family/arm/m3/Hwi_asm*.sv7M
- Zero-latency interrupts in TI-RTOS (bypassing the kernel) and the no-RTOS examples in the SDK are affected by this erratum.

Workarounds: **Software not using the Memory Protection Unit (MPU):**

For software not using the Memory Protection Unit (MPU), the workaround can be to disable CPU write buffering (register CPU_SCS.ACTLR.DISDEFWBUF) at the cost of significantly reduced execution speed.

All other cases (recommended workaround):

Ensure a DSB instruction occurs between the store and the BX instruction. For exception handlers written in C, this can be achieved by inserting the appropriate set of intrinsics or inline assembly just before the end of the interrupt function, for example:

ARMCC:

```
...  
__schedule_barrier(); __asm{DSB}; __schedule_barrier(); }
```

GCC:

```
...  
__asm volatile ("dsb 0xf" ::: "memory"); }
```

Note

The workaround for this bug will **not** be added automatically by the compiler.

CPU_02**Arm® Errata #752770: Interrupted loads to SP can cause erroneous behavior****Revisions Affected:** Revision E and earlier**Details:**

An interrupt occurring during the data-phase of a single word load to the stack-pointer (SP/R13) can cause an erroneous behavior of the device. In all cases, returning from the interrupt will result in the load instruction being executed an additional time. For all instructions performing an update to the base register, the base register will be erroneously updated on each execution, resulting in the stack-pointer being loaded from an incorrect memory location.

The affected instructions that can result in the load transaction being repeated are:

- LDR SP,[Rn],#imm
- LDR SP,[Rn,#imm]!
- LDR SP,[Rn,#imm]
- LDR SP,[Rn]
- LDR SP,[Rn,Rm]

The affected instructions that can result in the stack-pointer being loaded from an incorrect memory address are:

- LDR SP,[Rn],#imm
- LDR SP,[Rn,#imm]!

Conditions:

- An LDR is executed, with SP/R13 as the destination.
- The address for the LDR is successfully issued to the memory system.
- An interrupt is taken before the data has been returned and written to the stack-pointer.

Implications:

Unless the load is being performed to device memory or strongly-ordered memory, there should be no implications from the repetition of the load.

- In the unlikely event that the load is being performed to device memory or strongly-ordered memory, the repeated read can result in the final stack-pointer value being different than had only a single load been performed.
- Interruption of the two write-back forms of the instruction can result in both the base register value and the final stack-pointer value being incorrect. This can result in apparent stack corruption and subsequent unintended modification of memory.

Workaround:

Most compilers ensure this bug is not triggered by not emitting the affected instruction sequence and not using the instructions in the compiler runtime libraries. This includes:

- IAR from v6.21
- All versions of TI's Arm compiler (CCS)

A workaround for both issues can be implemented by replacing the direct load to the stack-pointer, with an intermediate load to a general-purpose register followed by a move to the stack-pointer.

If repeated reads are acceptable, then the base register update issue may be worked around by performing the stack-pointer load without the base increment followed by a subsequent ADD or SUB instruction to perform the appropriate update to the base register.

CPU_03

Arm® Errata #776924: VDIV or VSQRT instructions might not complete correctly when very short ISRs are used

Revisions Affected: Revision E and earlier

Details:

On the Arm® Cortex®-M4F processor, the VDIV and VSQRT instructions take 14 cycles to execute. When an interrupt is taken a VDIV or VSQRT instruction is not terminated, and completes its execution while the interrupt stacking occurs. If lazy context save of floating point state is enabled then the automatic stacking of the floating point context does not occur until a floating point instruction is executed inside the interrupt service routine.

Lazy context save is enabled by default. When it is enabled, the minimum time for the first instruction in the interrupt service routine to start executing is 12 cycles. In certain timing conditions, and if there is only one or two instructions inside the interrupt service routine, then the VDIV or VSQRT instruction might not write its result to the register bank or to the FPSCR.

Conditions::

- The floating point unit is present and enabled
- Lazy context saving is not disabled
- A VDIV or VSQRT is executed
- The destination register for the VDIV or VSQRT is one of s0 - s15
- An interrupt occurs and is taken.
- The interrupt service routine being executed does not contain a floating point instruction.
- An interrupt return is executed 14 cycles after the VDIV or VSQRT is executed.

A minimum of 12 of these 14 cycles are utilized for the context state stacking, which leaves 2 cycles for instructions inside the interrupt service routine, or 2 wait states applied to the entire stacking sequence (which means that it is not a constant wait state for every access).

In general this means that if the memory system inserts wait states for stack transactions then this erratum cannot be observed.

Implications:

The VDIV or VSQRT instruction does not complete correctly and the register bank and FPSCR are not updated, meaning that these registers hold incorrect, out of date, data.

For hand-written assembly code inside interrupt routines, this erratum should be considered.

Workarounds:

A workaround is only required if the floating point unit is present and enabled. A workaround is not required if the memory system inserts one or more wait states to every stack transaction.

When using TI-RTOS interrupts, all interrupt service routines will contain more than the 2 instructions and no workaround is required.

In all other cases, one of the following two workarounds must be implemented:

Workaround 1: Disable lazy context save of floating point state by clearing LSPEN to 0 (bit 30 of the FPCCR at address 0xE000EF34).

Workaround 2: Ensure that every interrupt service routine contains more than 2 instructions in addition to the exception return instruction.

CPU_Sys_01 *The SysTick calibration value (register field CPU_SCS.STCR.TENMS) used to set up 10-ms periodic ticks is incorrect when the system CPU is running off divided down 48-MHz clock*

Revisions Affected: Revision E and earlier

Details: When using the Arm® Cortex® SysTick timer, the TENMS register field (CPU_SCS.STCR.TENMS) will always show the value corresponding to a 48-MHz CPU clock, regardless of the CPU division factor.

Workarounds: One of the following two workarounds must be implemented:

Workaround 1: Do not use a divided down system CPU clock. In general, power savings are maximized by completing a task at full clock speed and then stopping the system CPU entirely after the task is complete.

Workaround 2: Read the system CPU division factor from the PRCM.CPUCLKDIV.RATIO register and compensate the TENMS field in software based on this value.

TI-provided drivers do not offer any functionality to divide the system CPU clock.

Sys_01 *Device might boot into ROM serial bootloader when waking up from shutdown*

Revisions Affected: Revision E and earlier

Details: For the conditions given below, the device will boot into and execute the ROM serial bootloader when waking up from Shutdown power mode. Intended behavior is to execute the application image. The prerequisites for this erratum to happen are:

- The wake up from Shutdown must be caused by toggling or noise on the JTAG TCK pin and not by a GPIO event.
- The Customer Configuration Section (CCFG) must have configured the bootloader with the following field values:
 - BOOTLOADER_ENABLE = 0xC5 (Bootloader enabled)
 - BL_ENABLE = 0xC5 (Bootloader pin backdoor enabled)
 - BL_PIN_NUMBER = n (any valid DIO number)

With the above prerequisites, the bootloader will be entered in the following cases:

- The CCFG bootloader pin level (BL_LEVEL) is set to 0x0 (active low) AND the input buffer enable for the DIO defined in BL_PIN_NUMBER is disabled in register IOC.IOCFGn.IE. If the input buffer is not enabled, the DIO level will always read 0 and bootloader will be entered.
- The input buffer controlled by IOC.IOCFGn.IE is enabled and the DIO input value is the same level as the CCFG bootloader pin level (BL_LEVEL) when entering Shutdown (GPIO input values are latched when entering Shutdown)

Please refer to the ICeMelter chapter in the [CC13x2, CC26x2 SimpleLink™ Wireless MCU Technical Reference Manual](#) for details on how noise entering the JTAG TCK pin can wake up the device

Workarounds: One of the following workarounds must be implemented:

- If input buffer is not enabled, use only active high bootloader pin level (BL_LEVEL)
- If input buffer is enabled, ensure DIO input pin level is not the same as bootloader pin level (BL_LEVEL) when entering Shutdown.

SYCTRL_01 *Resets occurring in a specific 2-MHz period during initial power up are incorrectly reported*

Revisions Affected: Revision E and earlier

Details: If a reset occurs in a specific 2-MHz period during initial power-up (boot), the reset source in AON_PMCTL.RESETCTL.RESET_SRC is reported as PWR_ON regardless of the reset source. This means that there is a window of 0.5 μ s during boot where a reset can be incorrectly reported.

Workaround: None

IOC_01***Limited number of DIOs available for the bootloader backdoor***

Revisions Affected: Revision E and earlier**Details:** The highest possible DIO number that can be used for the bootloader backdoor is limited to the number of available GPIOs minus 1. The bootloader backdoor pin is configured through SET_CCFG_BL_CONFIG_BL_PIN_NUMBER in ccfg.c. That means that if the device has x GPIOs, the highest DIO number that can be selected for the bootloader backdoor is DIO_{x-1}, even if higher DIO numbers are available for the device.**Workarounds:** There are no workarounds for this issue.

SRAM_01 ***Reserved addresses within SRAM_MMR region alias into SRAM array***

Revisions Affected: Revision E and earlier

Details: Accessing addresses within SRAM_MMR and greater than SRAM_MMR.MEM_CTL (0x40035010 - 0x4003FFFC) will alias into the SRAM array (0x20000010 - 0x20000FFFC) for both reading and writing.

Workarounds: Avoid accessing reserved addresses within SRAM_MMR.

Note that if using the Memory Protection Unit (MPU) to protect accesses to the SRAM array, also consider protecting the reserved SRAM_MMR addresses.

GPTM_01 ***An Incorrect Value Might Be Written to the General-Purpose (GP) Timers MMRs (Memory Mapped Registers) When Simultaneously Accessing the PKA (Public Key Accelerator) Engine and/or the AES (Advanced Encryption Standard) Engine from a Different Master***

Revisions Affected: Revision E and earlier

Details: When writing data to the GP Timer MMRs from one master while simultaneously accessing the PKA/AES modules from another master (read/write), an incorrect value might be captured in the GP Timer MMRs. In some cases, the incorrect value is replaced by the correct one after two clock cycles, but not always. No issue is seen when accessing the modules from the same master.

Workaround 1: Avoid accesses to PKA/AES by other masters while writing to the GP Timer MMRs. This can be accomplished by acquiring the relevant semaphores (depending on drivers/stacks) before writing to the GP Timer.

Note that the BLE stack executes crypto operations from ISR context. Other software must release the semaphores from a higher-priority ISR to avoid deadlocks.

Workaround 2: Verify the value written to the GP Timer MMR by reading it back in software. Correct the value if necessary.

ADC_01 ***Periodic ADC trigger at 200 kHz rate can be ignored when XOSC_HF is turned on or off***

Revisions Affected: Revision E and earlier

Details: There is no dedicated clock source selection for the ADC clock. The clock is derived from either XOSC_HF or RCOSC_HF, but defaults to XOSC_HF-derived clock whenever this is turned on.

When the ADC clock source is switched from RCOSC_HF to XOSC_HF-derived clock, the clock will stop for 2 cycles (24 MHz).

When the ADC clock source is switched from XOSC_HF-derived clock to RCOSC_HF-derived clock, the clock will stop for additionally 12 clock cycles, as the RCOSC_HF-derived clock is not ready when switch is done.

The fact that the clock is stopped, together with the difference in frequency between XOSC_HF and RCOSC_HF, may cause the ADC sampling and conversion to finish too late to catch the next trigger.

Workaround 1: Use asynchronous sampling.

The sampling period after the issue occurs can be reduced by up to 20% (12 + 1 clock cycles at 24 MHz)

To use the ADC in asynchronous mode from the Sensor Controller:

Call `adcEnableAsync()` to enable the ADC, instead of `adcEnableSync()`

Example:

```
adcEnableAsync(ADC_REF_FIXED, ADC_TRIGGER_AUX_TIMER0);
```

To use the ADC in asynchronous mode from the System CPU, by using the ADCBuf driver:

```
ADCBuf_Params params;
ADCBufCC26X2_ParamsExtension paramsExtension;

ADCBuf_Params_init(&params);
ADCBufCC26X2_ParamsExtension_init(&paramsExtension);

paramsExtension.samplingMode = ADCBufCC26X2_SAMPING_MODE_ASYNCHRONOUS;
params.custom = &paramsExtension;
```

To use the ADC in asynchronous mode from the System CPU, by using DriverLib API:

Call `AUXADCEnableAsync()` to enable the ADC, instead of `AUXADCEnableSync()`

Example:

```
AUXADCEnableAsync(AUXADC_REF_FIXED, AUXADC_TRIGGER_GPT0A);
```

Please note the difference between the asynchronous and synchronous ADC modes:

- In asynchronous mode, the ADC trigger ends the sampling period (which started immediately after the previous conversion), and starts conversion.
- In synchronous mode, the ADC trigger starts the sampling period (with configurable duration), followed by conversion.

Workaround 2: Ensure that XOSC_HF is not turned on or off while the ADC is used.

Workaround 3: Increase the sampling period by $(12+1)/24 \mu\text{s}$ or more.

ADC_02

ADC samples can be delayed by 2 or 14 clock cycles (24 MHz) when XOSC_HF is turned on or off, resulting in sample jitter

Revisions Affected: Revision E and earlier

Details:

There is no dedicated clock source selection for the ADC clock. The clock is derived from either XOSC_HF or RCOSC_HF, but defaults to XOSC_HF-derived clock whenever this is turned on.

When the ADC clock source is switched from RCOSC_HF to XOSC_HF-derived clock, the clock will stop for 2 cycles (24 MHz).

When the ADC clock source is switched from XOSC_HF-derived clock to RCOSC_HF-derived clock, the clock will stop for additionally 12 clock cycles, as the RCOSC_HF-derived clock is not ready when switch is done.

SCLK_HF switches from RCOSC_HF to XOSC_HF at different times compared to ADC clock. This leads to sample jitter.

Workaround 1:

Use asynchronous sampling

- This will reduce the delay of 14 clock cycles down to 2 clock cycles.
- Using asynchronous sampling and an external trigger source (GPIO input pin) will eliminate the delay completely

To use the ADC in asynchronous mode from the Sensor Controller:

Call `adcEnableAsync()` to enable the ADC, instead of `adcEnableSync()`

Example:

```
adcEnableAsync(ADC_REF_FIXED, ADC_TRIGGER_AUX_TIMER0);
```

To use the ADC in asynchronous mode from the System CPU, by using the ADCBuf driver:

```
ADCBuf_Params params;  
ADCBufCC26X2_ParamsExtension paramsExtension;  
  
ADCBuf_Params_init(&params);  
ADCBufCC26X2_ParamsExtension_init(&paramsExtension);  
  
paramsExtension.samplingMode = ADCBufCC26X2_SAMPING_MODE_ASYNCHRONOUS;  
params.custom = &paramsExtension;
```

To use the ADC in asynchronous mode from the System CPU, by using DriverLib API:

Call `AUXADCEnableAsync()` to enable the ADC, instead of `AUXADCEnableSync()`

Example:

```
AUXADCEnableAsync(AUXADC_REF_FIXED, AUXADC_TRIGGER_GPT0A);
```

Please note the difference between the asynchronous and synchronous ADC modes:

- In asynchronous mode, the ADC trigger ends the sampling period (which started immediately after the previous conversion), and starts conversion.
- In synchronous mode, the ADC trigger starts the sampling period (with configurable duration), followed by conversion

Workaround 2:

Ensure that XOSC_HF is not turned on or off while the ADC is used.

ADC_03

Software can hang when reading the ADC FIFO if a single manual ADC trigger is generated immediately after the ADC is enabled

Revisions Affected: Revision E and earlier

Details:

There is no dedicated clock source selection for the ADC clock. The clock is derived from either XOSC_HF or RCOSC_HF, but defaults to XOSC_HF-derived clock whenever this is turned on.

When the ADC clock source is switched from RCOSC_HF to XOSC_HF-derived clock, the clock will stop for 2 cycles (24 MHz).

When the ADC clock source is switched from XOSC_HF-derived clock to RCOSC_HF-derived clock, the clock will stop for additionally 12 clock cycles, as the RCOSC_HF-derived clock is not ready when switch is done.

The additional 12 clock cycles introduces a race between trigger-event and ADC trigger-detector to get out of reset.

Workaround 1: Updated TI software adds a short delay at the end of the function that enables the ADC.

- If using the ADC through the System CPU (TI drivers or DriverLib API): Use SimpleLink CC13x2 and CC26x2 SDK 4.30 or later (release end of Q3 2020)
- If using ADC through the Sensor Controller (ADC resource): Use Sensor Controller Studio 2.7.0 or later (release end of Q2 2020), or use the update service to install a patch for Sensor Controller Studio 2.0.0 through 2.6.0

Workaround 2: Ensure that XOSC_HF is not turned on or off while the ADC is used.

ADC_04

Misbehaving ADC FIFO status flags in the AUX_ANAIF:ADCFIFOSTAT register (OVERFLOW, FULL, ALMOST_FULL, and EMPTY)

Revisions Affected: Revision E and earlier

Details: The ADC FIFO status flags do not behave correctly when the ADC FIFO is being read at the same clock cycle as a new sample is added.

Scenario A:

The EMPTY flag goes high even if there is one sample left in the ADC FIFO. This can happen if there is one sample in the ADC FIFO (the EMPTY flag is low), and this sample is being read on the same clock cycle as the ADC adds another sample. In this case, the EMPTY flag is incorrectly set high, and reading the ADC FIFO will cause an underflow even though the ADC FIFO is not empty. The only way to recover without a flush (AUX_ANAIF:ADCCTL.CMD = 3) is to add another sample (ADC or debug write).

Scenario B:

The ALMOST_FULL flag goes low when there are three samples in the ADC FIFO. This can happen if there are three samples in the ADC FIFO (the ALMOST_FULL flag is high), and one of the samples is being read on the same clock cycle as the ADC adds another sample. In this case, the ALMOST_FULL flag is incorrectly set low, even if there are still 3 new samples in the ADC FIFO.

Scenario C:

The FULL flag goes low while the ALMOST_FULL flag and OVERFLOW flag go high when there are four samples in the ADC FIFO. This can happen if there are four samples in the ADC FIFO (the FIFO_FULL flag is high) and one of the samples is being read on the same clock cycle as the ADC adds another sample. In this case, the FULL flag is incorrectly set low and the ALMOST_FULL flag and OVERFLOW flag are set high.

Workaround: The CPU must use the ADC FIFO flags or corresponding events to know when to read the samples from the ADC FIFO.

The scenarios described will not occur if the CPU processing time from sample notification to completed ADC FIFO read is less than the ADC sampling period. Hence, once notified about available sample(s), the CPU should always read the ADC FIFO before starting other processing tasks. If the CPU processing time from sample notification to completed ADC FIFO read is longer than the ADC sampling period, the ADC FIFO will correctly overflow.

When using the Sensor Controller, you need to ensure that `adcReadFifo()` or `adcPopFifo()` is not called when an ADC conversion finishes. For most applications it makes sense to read the ADC FIFO before another sample is added to the FIFO, so that the FIFO never contains more than one sample, and only contains one sample for a short time.

In practice, this means:

- When using manual trigger: Call `adcReadFifo()` to empty the FIFO before or immediately after generating another manual trigger
- When using timer trigger: Call `adcReadFifo()` to empty the FIFO before or immediately after generating another timer trigger. As this is typically implemented with a loop (with one iteration per ADC sample), ensure that the worst-case loop iteration time is less than one ADC sample period.
- When using other triggers, which may or may not occur: Use `adcGetFifoState()` to poll the ADC FIFO state, and use `adcPopFifo()` to empty the FIFO as quickly as possible. Also, before disabling the ADC, ensure that no more triggers can occur (for

example by disabling a GPIO input buffer) and wait until any ongoing ADC operation has finished.

ADC_05 ***Writing any value to AUX_ANAIF:ADCTRIG.START will create an ADC trigger***

Revisions Affected: Revision E and earlier

Details: The original register documentation states that you need to write a '1' to AUX_ANAIF:ADCTRIG.START to manually trigger the ADC. This is not correct, as any write will do this.

Workaround: Avoid writing to AUX_ANAIF:ADCTRIG.START unless an ADC trigger should be generated.

4 Revision History

Changes from May 24, 2019 to December 2, 2020 (from Revision C (May 2019) to Revision D (December 2020))

Page

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- Added the following advisories: IOC_01, SRAM_01, GPTM_01, ADC_01, ADC_02, ADC_03, ADC_04, ADC_05..... [2](#)
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